

# Ultrasonic Characterization on Sequences of CFRP Composites Based on Modeling and Motorized System

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Composites are a material class for which nondestructive material property characterization is as important as flaw detection. Laminates of fiber reinforced composites often possess strong in-plane elastic anisotropy attributable to the specific fiber orientation and layup sequence when waves are propagating in the thickness direction of composite laminates. So the layup orientation greatly influences its properties in a composite laminate. It could result in the part being rejected and discarded if the layup orientation of a ply is misaligned. A nondestructive technique would be very beneficial, which could be used to test the part after curing and requires less time than the optical test. Therefore a ply-by-ply vector decomposition model has been developed, simplified, and implemented for composite laminates fabricated from unidirectional plies. This model decomposes the transmission of a linearly polarized ultrasound wave into orthogonal components through each ply of a laminate. Also in order to develop these methods into practical inspection tools, motorized system have been developed for different measurement modalities for acquiring ultrasonic signals as a function of in-plane angle. It is found that high probability shows between the model and tests developed in characterizing cured layups of the laminates.

**Key Words :** Sequences, Composite Laminates, Ply-by-ply Vector

## 1. Introduction

Composite materials (Im et al., 2002) are attractive for a wide range of applications because of their high strength, stiffness and low density, composites are currently being considered for many structural (aerospace vehicles, automobiles,

trains and ships) applications due to their potential for reducing structural weight. Increasingly more high performance engineering structures are being built with critical structural components made from composite materials. Especially, the importance of carbon-fiber reinforced plastics (CFRP) has been generally recognized, and CFRP composite laminates are widely used. So, CFRPs are a material class for which non-destructive material property characterization is as important as flaw detection (Hsu, 1994; Fisher, 1996; Hsu and Margetan, 1993; Hsu et al., 1997; Hale et al., 1996). Fiber reinforced composite laminates often possess strong in-plane

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elastic anisotropy attributable to the specific fiber orientation and layup sequence (Tippler, 1982). However one of important factors is the layup sequence which can influence the CFRP composite performance. This greatly affects its properties in the composite laminate. If one ply is misaligned in the layup sequence, it can drastically alter the mechanical performance of composite laminates. So most manufacturers cut a small sample from the waste edge and use a microscope to optically verify the ply orientations on critical parts, which adds more cost to the composites due to intensive labor and performance after curing.

Many of these elastic anisotropies may be investigated using ultrasound, among which angular measurements are often used. Hsu (1994) used angular scan of acousto-ultrasonic signals to investigate fiber reinforced composite laminates. By placing and rotating two contact transducers on the same side of crossed-ply composite laminates, the angular dependence of the acousto-ultrasonic signal was measured and found to have good correlation with the fiber orientation of the sample. Angular measurement of normal-incident shear wave has also been used to detect errors in layup sequence and ply orientation in both green (before cured) and cured composites. The transmitted signals of normal incident shear wave in a "crossed polarizer" configuration were found to be particularly sensitive to ply misorientation and layup sequence in a laminate. Also Urabe and Yomoda (1982, 1987) have utilized a non-destructive method using a 4 GHz microwave to determine the fiber orientation in the CFRP composites. This method is based on the electrical anisotropy in the orthotropic directions of a unidirectional laminate, with the principal direction aligned with the direction of the fibers. For this method, an incident standing wave is projected into the sample at a given orientation. The differences between the received signals, one with the receiving unit polarized horizontally and one with the receiving unit polarized vertically with respect to the apparatus, were used to determine the fiber orientation in samples constructed from two to eight plies of prepreg tape. Also, Urabe (Urabe, 1987) conducted a research using

a 35 GHz microwave to determine fiber orientation in carbon fiber reinforced plastics. Studies were performed on thick composite laminates by Komsky et al. (1994). So, they have developed a method and successfully used it to predict the layer orientation for a 70-ply layer laminate through a neural network. And Komsky et al. (1992) have researched the interaction of ultrasonic shear waves with thick composite laminates. This research studied the transmission of shear waves through laminates with both the transmitter and receiver aligned and fixed with respect to the fiber directions in the outer layers of the sample. Results of the experiments showed that changing the layer orientations in the laminate caused a direct effect on the received signal shape and amplitude. A layer-by-layer vector decomposition model was presented to theoretically explain the interaction of shear waves as they propagated through the laminate, but no simulated results were given for the samples tested.

This paper aims at developing a motorized scanner for different measurement modalities for acquiring ultrasonic signals as a function of in-plane angle. The system is a motorized, PC-controlled angular scanner for making transmission measurements using a pair of normal-incidence shear wave transducers and investigating the inspection of ply misorientation and layup sequence in CFRP laminates when normal incident ultrasound waves transmitted in the thickness direction of composite laminates. Therefore, a motorized scan system and a new technique are presented for determining ply orientation errors and sequencing errors in a composite laminate using through transmission of shear waves based on the theoretical ply-by-ply vector decomposition.

## 2. Theoretical Approach

Hsu et al. (1997) suggested that this model decomposes the transmission of a linearly polarized ultrasound wave into orthogonal components through each ply of a laminate. The input to the first ply is decomposed into one component which propagates through the first ply

parallel to the fibers and one component which propagates through the first ply perpendicular to the fibers. These two components then become the input for the second ply, where each one is then decomposed into components parallel and perpendicular to the fibers in the second ply. This process continues for all remaining plies in the laminate. After all of the components have been decomposed and have propagated through the last ply in the laminate, the components are projected onto the axis of the receiver and are combined to predict a received signal. The model is general and applies to any layup produced from plies with orthogonal properties.

In the derivation of the model,  $\alpha_i$ ,  $\beta_i$ , and  $\rho$  are the attenuation coefficients of the orientations of the transmitter, the waves polarized parallel to the fibers and the couplant layers,  $(h_i)$  is the thickness of the  $i^{\text{th}}$  ply, and  $(tt)$  and  $(tr)$  are the couplant thicknesses at the transmitter and receiver, respectively. Losses due to beam spreading and interface losses, which depend on the ply to ply orientations and thicknesses, are included in the  $f_{ij}(\delta)$  signal reduction factor. The ply-by-ply vector decomposition model can now be derived.

A wave pulse,  $S_T$ , is generated by the transmitter at a transmitter angle  $\alpha_T$  and propagates through the couplant to the face of the first ply. The wave now has the transmitter amplitude  $S_T e^{-\rho(tt)}$  due to the signal attenuation of the couplant, and a time shift equal to the thickness of the couplant divided by the wave velocity through the couplant. This signal is then decomposed into two components through the change angle of each interface  $\Delta\theta_1 = \alpha_T - \alpha_1$  in directions parallel and perpendicular to the fibers in the first ply. These two components shown in Fig. 1 then propagate through the first ply and are reduced by their respective attenuation, interface, and beam spreading losses. Time shifts of the two components, caused by the fact that the wave velocities,  $\nu_a$  and  $\nu_b$  are not equal, are tabulated along with the magnitude changes due to losses in lieu of the typical phase term  $e^{-ikx}$ . New values and time shifts for each component are now given by Eq. (1).

Parallel Component in the first ply at the first term of  $\Delta t = (tt)/\nu_\rho + (h1)/\nu_a$  :

$$S_{1para.} = S_T e^{-\rho(tt)} e^{-\alpha_1(h1)} f_{11}(\delta) \cos(\Delta\theta_1) \tag{1}$$

Perpendicular Component in the first ply at the first term of  $\Delta t = (tt)/\nu_\rho + (h1)/\nu_b$  :

$$S_{1perp.} = S_T e^{-\rho(tt)} e^{-\beta_1(h1)} f_{12}(\delta) \sin(\Delta\theta_1)$$

where  $e^{-\alpha_1(h1)}$  is the attenuation for the waves polarized parallel to the fibers in the  $i^{\text{th}}$  ply,  $e^{-\beta_1(h1)}$  is the attenuation for the waves polarized perpendicular to the fibers in the  $i^{\text{th}}$  ply,  $e^{-\rho(tt)}$  is the signal attenuations in the couplant layers,  $e^{-\rho(tr)}$  is the signal attenuations in the couplant layers, and  $\nu_a$ ,  $\nu_b$  and  $\nu_\rho$  are the wave velocities when the component is polarized parallel to the fibers, perpendicular to the fibers, and propagating through the couplant respectively.

As these two components enter the second ply, with the fibers oriented at  $\alpha_2$ , each one will contribute two more components if angle  $\alpha_2$  does not equal angle  $\alpha_1$ . The components are projected onto the second ply, reduced by losses, and experience another time shift as they propagate into and through the second ply, given by Eq. (2). Note that each component must retain its corresponding time shift through the calculation as shown in Fig. 2.

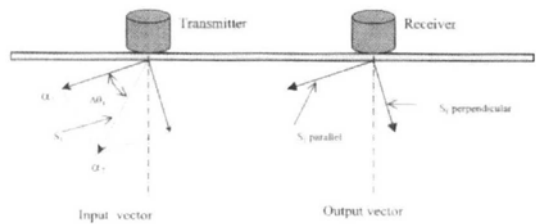


Fig. 1 Vector projection in case of input and output

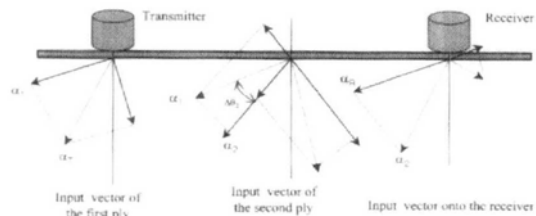


Fig. 2 Schematic of the ply to ply vector projections

Parallel Components in 2nd ply at the first term of  $\Delta t = (tt)/\nu_\rho + (h1)/\nu_\alpha + (h2)/\nu_\alpha$  :

$$S_{2para} = S_T e^{-\rho(tt)} * \left\{ e^{-\alpha_1(h1)} e^{-2(h2)} f_{11}(\delta) f_{21}(\delta) \cos(\Delta\theta_1) \cos(\Delta\theta_2) \right. \\ \left. + e^{-\beta_1(h1)} e^{-\alpha_2(h2)} f_{12}(\delta) f_{24}(\delta) \sin(\Delta\theta_1) \sin(\Delta\theta_2) \right\} \quad (2)$$

Perpendicular Components in 2nd ply at the first term of  $\Delta t = (tt)/\nu_\rho + (h1)/\nu_\beta + (h2)/\nu_\beta$  :

$$S_{2perp} = S_T e^{-\rho(tt)} * \left\{ e^{-\beta_1(h1)} e^{-\beta_2(h2)} f_{12}(\delta) f_{22}(\delta) \sin(\Delta\theta_1) \cos(\Delta\theta_2) \right. \\ \left. - e^{-\alpha_1(h1)} e^{-2(h2)} f_{11}(\delta) f_{23}(\delta) \cos(\Delta\theta_1) \sin(\Delta\theta_2) \right\}$$

This process continues in the same manner for subsequent plies in the plate at an exponential rate of  $2^i$  components decomposed for the  $i^{\text{th}}$  ply, and a plate with  $N$  plies will have a total of  $2^N$  discrete, decomposed components. To demonstrate the projection of the signals onto the receiver, a plate with only two plies will be used. The signal components exiting in the second ply are given in Eq. (2) and all four components will be reduced by  $e^{-\rho(tr)}$  as they propagate through the couplant to the face of the receiver. The parallel component is then projected onto the receiver axis by multiplying it by  $\cos(\Delta\theta_R)$ ; likewise, the perpendicular component is projected onto the receiver axis by multiplying it by  $\sin(\Delta\theta_R)$ , where  $\Delta\theta_R = \alpha_R - \alpha_2$ . The received signal is given by Eq. (3). This equation demonstrates that the received signal is the combination of four new time varying vectors for a plate with two plies. Each new vector is constructed by multiplying the original time varying vector,  $S_T$ , by a scalar term, and shifting the new vector by its corresponding  $\Delta t$ .

Component parallel to receiver axis at the first term of  $\Delta t = (tt)/\nu_\rho + (h1)/\nu_\alpha + (h2)/\nu_\rho + (tr)/\nu_\rho$  :

$$S_R = S_T e^{-\rho(tt)} e^{-\rho(tr)} * \left\{ e^{-\alpha_1(h1)} e^{-\alpha_2(h2)} f_{11}(\delta) f_{22}(\delta) \cos(\Delta\theta_1) \cos(\Delta\theta_2) \cos(\Delta\theta_R) \right. \\ \left. + e^{-\beta_1(h1)} e^{-2(h2)} f_{12}(\delta) f_{24}(\delta) \sin(\Delta\theta_1) \sin(\Delta\theta_2) \cos(\Delta\theta_R) \right. \\ \left. - e^{-\alpha_1(h1)} e^{-\beta_2(h2)} f_{11}(\delta) f_{23}(\delta) \cos(\Delta\theta_1) \sin(\Delta\theta_2) \sin(\Delta\theta_R) \right. \\ \left. + e^{-\beta_1(h1)} e^{-\beta_2(h2)} f_{12}(\delta) f_{22}(\delta) \sin(\Delta\theta_1) \cos(\Delta\theta_2) \sin(\Delta\theta_R) \right\} \quad (3)$$

The ply-by-ply vector decomposition model is a very powerful tool for the ultrasonic testing of composite plates consisting of unidirectional plies. As with most theoretical methods, the ply-by-ply vector decomposition model has many complications involved with its implementation and requires that some assumptions be made to

reduce it to a practical level for nondestructive testing.

### 3. Simplified Model

The final form for the theoretical ply-by-ply vector decomposition model, as well as similar decomposition models (Hsu et al., 1997), for a typical plate with numerous plies is a very lengthy and complex calculation. As shown previously, the number of discrete components in the theoretical model grow at an exponential rate of  $2^N$ . Thus, a plate consisting of 40 plies will contribute  $2^{40}$ , approximately 1.995 trillion, discrete terms to the calculation of  $S_R$ . An array of this size would be a highly time-consuming and inefficient computation even with the speed and power of today's computers. Another complication arises in determining the values and functions of  $\alpha$ ,  $\beta$ ,  $\rho$ , and  $f_{ij}(\delta)$  to be used in the computation of  $S_R$  for many plies. In order to utilize the ply-by-ply vector decomposition technique, some assumptions and simplifications are made to simplify the theoretical model.

A plate consisting of two plies will be used to illustrate the transformation of the theoretical model to the simplified ply-by-ply vector decomposition model. The first assumption used is that the interface and beam spreading losses are negligible, that is, all  $f_{ij}(\delta)$  terms equal one. This may seem to be a drastic assumption since these losses are sensitive to the order and orientation of the plies, but preliminary tests have shown this assumption to have little effect on the qualitative results and it is essential for further simplification. Next, by constraining all plies in the plate to be comprised of the same material,  $(h_1)$ ,  $\alpha_1$ , and  $\beta_1$  are assumed to be identical for every ply. Finally, the couplant is a thin layer of highly viscous material and when the transducers are pressed onto the plate, it is assumed that the couplant spreads into a uniform and equal thickness for both transducers; therefore,  $\rho(tt) = \rho(tr) = \rho'$ . The result of incorporating these assumptions into Eq. (3) and grouping like terms is given by Eq. (4) at the first term of  $\Delta t = (2tt)/\nu_\rho + (2h)$  ( $s_\alpha$ ).

$$\begin{aligned}
 S_R = & S_{Te}^{-2\rho'} e^{-2\alpha'} \cos(\Delta\theta_1) \cos(\Delta\theta_2) \cos(\Delta\theta_R) \\
 & + S_{Te}^{-2\rho'} e^{-\beta'} e^{-\alpha'} \sin(\Delta\theta_1) \sin(\Delta\theta_2) \cos(\Delta\theta_R) \\
 & - S_{Te}^{-2\rho'} e^{-\alpha'} e^{\beta'} \cos(\Delta\theta_1) \sin(\Delta\theta_2) \sin(\Delta\theta_R) \\
 & + S_{Te}^{-2\rho'} e^{-\beta'} \sin(\Delta\theta_1) \cos(\Delta\theta_2) \sin(\Delta\theta_R)
 \end{aligned} \quad (4)$$

where  $h = h_1, \dots, h_N$ ,  $\alpha' = \alpha_1(h_1), \dots, \alpha_N(h_N)$ ,  $\beta' = \beta_1(h_1), \dots, \beta_N(h_N)$ ,  $s_\alpha = 1/\nu_\alpha$ , and  $s_\beta = 1/\nu_\beta$ .

### 4. Experimental Method

CFRPs made from uni-directional prepreg sheets of carbon fibers (CU125NS) by Korea HANKUK Fiber Co., have the material properties shown in Table 1, based on the manufacturer’s specifications. The CFRP composite laminates are made of 40 plies of these sheets stacked at different angles. They are cured by heating to the appropriate hardening temperature (130°C) by a heater in the vaccum bag of the autoclave. Three types of specimens were used in this experimentation. Their lay-up, stacked with 40 plies, indicates that specimen A is [(0/45/90/-45)<sub>10</sub>]<sub>T</sub>, specimen B is [(0/45/90/-45)<sub>5</sub>]<sub>S</sub> and specimen C is [(0/45/90/-45)<sub>5</sub>]<sub>S</sub> with the 20<sup>th</sup> ply at +45° instead of -45°. Test specimens were prepared with dimensions, 70 mm × 70 mm × 5.23 mm (width × length × thickness). And the fiber-direction of specimen surface is manufactured to correspond to 0° direction; thus, the fiber-direction is the same as the length direction.

A schematic diagram of the motorized contact mode scanner is shown in Fig. 3. Two 1 MHz, 12.7 mm in diameter shear wave transducers in contact transmission mode, one of which serves as a transmitter and the other a receiver, had been

used. The two transducers are supported by a holder which can be rotated by a stepper motor with a maximum resolution of 0.9 degree. Each time after a full revolution of scan, the scanner reverses to the starting position so that the cables to the transducers will not continue to wrap around the axle. Two adjustable small springs are used to maintain a certain pressure on the transducers so that the transducers are kept in contact with the sample during the transmission scan mode. The burnt honey couplant can be used for the scan. The motorized system is assembled for experimental acquisition (see Fig. 4) by first placing the test specimen face down on the transmitter face plate and aligning its 90° axis between two transducers. A peak-to-peak amplitude was taken on the digital oscilloscope during data processing. The system consists of a PC, a pulser-receiver, a motor driver and the contact transmission mode probes described above. The parallel port of the PC is used to output the motor control signal and a high-speed data acquisition board is used to digitize the ultrasonic signals. A graphics based user interface software was developed for scan control and data analysis. With the motorized scanner, a typical transmiission

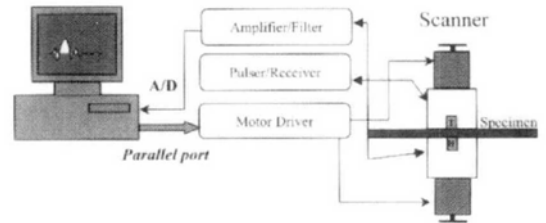


Fig. 3 Schematic for motorized scan system

**Table 1** Material properties

Characteristics	Fiber	Resin	Prepreg
Density	1.75 × 10 <sup>3</sup> [kg/m <sup>3</sup> ]	1.24 × 10 <sup>3</sup> [kg/m <sup>3</sup> ]	CU125NS
Tensile strength	3.53 [GPa]	0.078 [GPa]	
Elastic modulus	230 [GPa]	3.96 [GPa]	
Elongation	1.5 [%]	2.0 [%]	
Resin content			37 [% Wt]

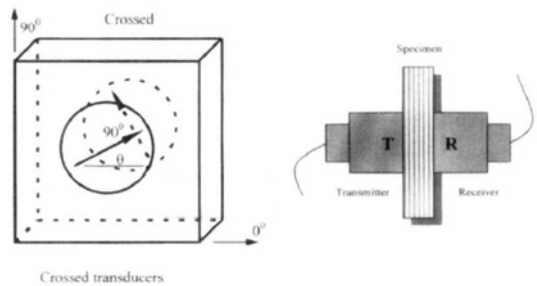


Fig. 4 Schematic of transducer contact position in through-transmission mode

scan (360 degree with a step of 1.8 degree) can be done in less than one minute. Here this instrumentation included a Panametrics #5075PR pulser/receiver and a LeCroy 9400 digital oscilloscope. The waves were generated and received using a pair of Panametrics VI53, 12.7 mm diameter, 1 MHz, shear wave transducers which were coupled to the composite laminates using a burnt honey couplant supplied by Panametrics.

### 5. Discussion and Results

#### 5.1 Verification of the simplified model

Experiments were performed on specimens fabricated from 40 plies of graphite/epoxy prepreg sheets, carbon fibers (CU125NS) by Korea HANKUK Fiber Co., with all fibers oriented parallel to each other to verify the simplified ply-by-ply vector decomposition model. All tests utilized a through-transmitted ultrasonic pulse which was generated and received by a pair of 1 MHz shear wave transducers. An impulse signal was generated by a pulser/receiver, and the received signal was displayed on a digital oscilloscope. To verify that the reduced model worked at various transducer angles, the first set of experiments were performed using one specimen of 40 aligned plies. This simplifies the ply-by-ply vector decomposition model to Eq. (5) below, where  $\Delta\theta_1 = \alpha_T - \alpha_1$ ,  $\Delta\theta_R = \alpha_R - \alpha_{40}$ , and  $R[0_{40}/90_0]$  and  $R[0_0/90_{40}]$  are the experimentally acquired signals. At that time, the other 39 reference signals are not required since all of the fibers in the pack are assumed to be perfectly aligned. These terms  $\sin(\Delta\theta) = 0$ , for  $i=2$  to 40, produce coefficients equal to zero for those reference signals.

$$S_R = R[0_{40}/90_0] \cos(\Delta\theta_1) \cos(\Delta\theta_R) + R[0_0/90_{40}] \sin(\Delta\theta_1) \sin(\Delta\theta_R) \quad (5)$$

The above signals were acquired and stored in the personal computer by aligning the transducers parallel to the fibers for  $R[0_{40}/90_0]$  and perpendicular to the fiber for  $R[0_0/90_{40}]$ . The transmitter and receiver were then oriented at specified angles,  $\alpha_T$  and  $\alpha_R$ , to the fibers and the received signal for each orientation was acquired

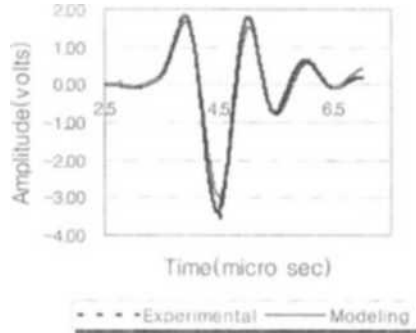


Fig. 5 Comparisons of modeling and experimental solutions

and stored in the computer. A spreadsheet was utilized to model the received signal from the stored reference signals and the transducer orientations using Eq. (5). Fig. 5 shows a comparison of the received signal and the modeled signal. The slight differences in amplitudes can be attributed to a change of couplant conditions and ply orientation errors due to small random layup errors. This good agreement between the experimental and modeled wave forms verifies the simplified model.

#### 5.2 Experimental solutions and theoretical modeling

Experimentation consists of performing a polar scan using a through transmission with the transducers in a crossed arrangement, that is, the receiver polarization is oriented at  $90^0$  the polarization of the transmitter. For an isotropic material, this test will produce a null, or zero received signal at any transmitter orientation. However, for a laminate consisting of orthotropic plies, this test is very sensitive to fiber orientation and ply sequence as shown by both the computer modeled and experimental results. The specimens used for this test are based on a realistic layup sequence used in manufacturing composite components and possible errors which can occur during fabrication. When the fixture is assembled for experimental data acquisition by first placing the test specimen face down on the transmitter face plate and aligning its  $0^0 - 90^0$  axis between two transducers, it is found that the results was very sensitive with the aligned angle of  $90^0$  be-

tween the transmitter and receiver in the through-transmission method.

Specimens A, B and C were compared to demonstrate the test's capability and sensitivity in determining a misoriented ply or an unsymmetrical layup with the same base sequence as a symmetrical layup with the experimental and theoretical results shown in Figs. 6, 7 and 8. First of all, in the case of specimen A with unsymmetric layups, experimental and theoretical solutions

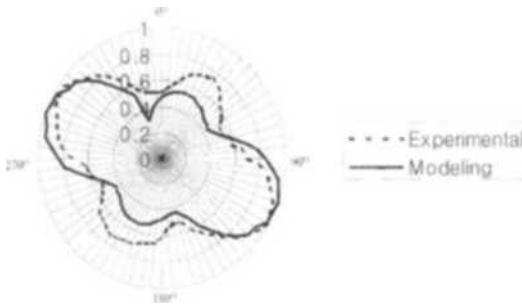


Fig. 6 Modeling and experimental solutions for specimen  $[(0/45/90/-45)_{10}]_T$

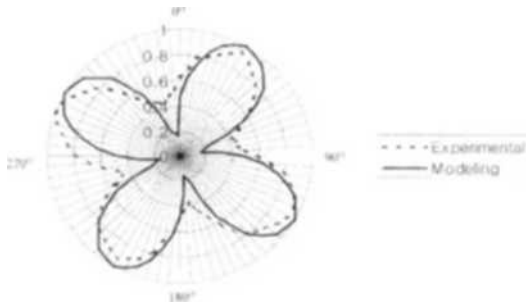


Fig. 7 Modeling and experimental solutions for specimen  $[(0/45/90/-45)_5]_S$

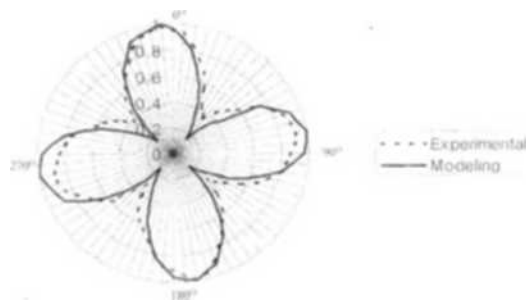


Fig. 8 Modeling and experimental solutions for specimen  $[(0/45/90/-45)_5]_S$

were obviously shown to some degree in Fig. 6. Figure 7 shows comparisons of experimental and theoretical results in the case of specimen B with symmetric layups. So it is found that there exists a good agreement in the amplitude and angle between experimental and theoretical results. Figure 8 shows comparisons of experimental and theoretical results in the case of specimen C with a single misoriented ply with symmetric layups. Again, a strong correlation is observed in this figure.

Experimental data was obtained by placing the transducers on each sample with a viscous couplant, orienting the transmitter, crossing the receiver's orientation such that a minimum peak-to-peak amplitude was displayed on the oscilloscope, and measuring the peak-to-peak amplitude of the minimized signal. This was done for transmitter orientations from  $0^\circ$  to  $360^\circ$  in  $5^\circ$  increments. The results for the experimental and modeling data are graphed in Figs 6, 7 and 8 for comparing difference between test and model. Modeling results were obtained by implementing the simplified ply-by-ply vector decomposition model. A program was written to compute the reference signal coefficients for any transducer and ply orientations, multiply the coefficients by their respective reference signal, combine the resultant signals to model a received signal, and compute the peak-to-peak amplitude of the synthesized signal. Since the dominant signal segments of  $R[0_{40}/90_0]$  and  $R[0_0/90_{40}]$  are approximately of the same shape and size, the required 41 reference signals were assembled by linearly interpolating the previously obtained  $R[0_{40}/90_0]$  reference signals through 40 time shifts. Each sample's layup sequence was entered, and the program computed the peak-to-peak values for the modeled signal for transmitter angles ranging from  $0^\circ$  to  $360^\circ$  in  $5^\circ$  increments with the receiver oriented at  $+90^\circ$  with respect to the transmitter. Those results show a strong qualitative agreement between the experimental and modeling solutions. The data in each plot was normalized by dividing it by the smallest peak-to-peak amplitude value contained in the data for each respective plot. The motorized system

had brought the effect of changing couplant conditions to be minimized from test to test which has a significant effect on the amplitude of the received signal. Also rapid scan experimentation can be made to find an error contained in the CFRP composite laminates.

## 6. Summary and Conclusions

A system has been used to determine the fiber orientation on graphite epoxy laminates, which is a motorized and PC-controlled ultrasonic scanner for contact mode in the shear wave transmission method using anisotropic composite plates. The system has been tested on unidirectional graphite epoxy laminate and a good agreement between the experiment and model was obtained. Especially, it was attempted to detect sequencing errors for symmetric and non-symmetric laminates about the mid plane, i.e.  $[(0/45/90/-45)_S]$  versus  $[(0/45/90/-45)_{10}_T]$ . To this end a through-transmission ultrasound test method has been performed to evaluate the layups of CFRP composite laminates. Both modeling and experimental results demonstrated high sensitivity of the method. In the case of a specimen with unsymmetric layups, a little difference was shown between experimental and modeling solutions. And there exists a good agreement between experimental and modeling results in the case of a specimen with symmetric layups. Also, a more strong correlation has been observed between experimental and modeling solutions for a specimen with a single misoriented ply with symmetric layups. This high sensitivity is good for characterizing the layup sequence in a laminate and detecting real manufacturing errors. And the implementation of the simplified ply-by-ply vector decomposition model has been successfully utilized to qualitatively model the behavior of ultrasound waves transmitted through a composite laminate fabricated from unidirectional plies. This model has the capability to qualitatively predict the effects of a ply misorientation and layup symmetry, as demonstrated by comparing the theoretical and experimental results. Also, ultrasound shear waves require a highly viscous

couplant between the transducer and the test piece. The usual shear wave couplant is burnt honey; unfortunately, it is very difficult to maintain a consistent property for burnt honey. Experimental measurement results could not be duplicated with a high degree of quantitative reproducibility due to the inconsistency of the couplant from test to test. However, the results could be reproduced to some degree because of the motorized system.

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